

TDC1147

Monolithic Video A/D Converter

7-Bit, 15 Msps

Description

The TDC1147 is a 7-bit "flash" analog-to-digital converter which has no pipeline delay between sampling and valid data. The output data register normally found on flash A/D converters has been bypassed, allowing data to transfer directly to output drivers from the encoding logic section of the circuit. The converter requires only one clock pulse to perform the complete conversion operation. The conversion time is guaranteed to be less than 60 nanoseconds.

The TDC1147 is function and pin-compatible with Raytheon Semiconductor La Jolla's TDC1047 7-bit flash A/D converter which has an output data register. The TDC1147 will operate accurately at sampling rates up to 15 Msps and has an analog bandwidth of 7 MHz. Linearity errors are guaranteed to be less than 0.4% over the operating temperature range.

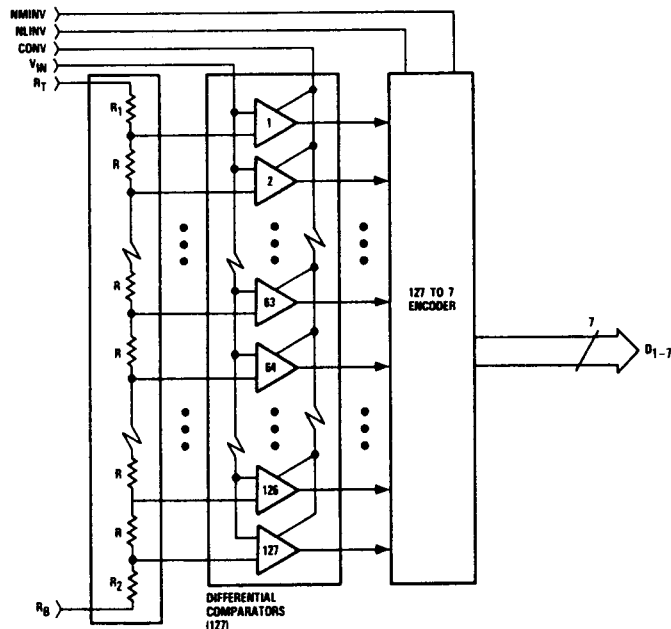
Features

- ◆ 20 Msps conversion rate
- ◆ No digital pipeline delay
- ◆ 7-bit resolution
- ◆ 1/2 LSB linearity
- ◆ Sample-and-hold circuit not required
- ◆ TTL compatible
- ◆ Selectable output format
- ◆ Available in 24 pin Cerdip

Applications

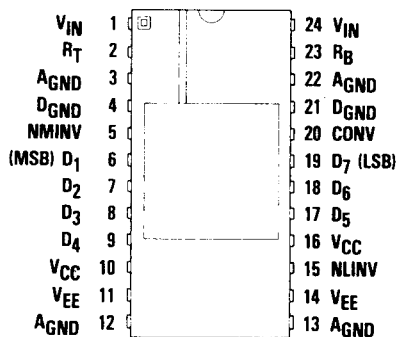
- ◆ Low-cost video digitizing
- ◆ Medical imaging
- ◆ Data acquisition
- ◆ High resolution A/D converters
- ◆ Telecommunications systems
- ◆ Radar data conversion

Functional Block Diagram



TDC1147

Pin Assignments



24 Pin CERDIP — B7 Package

and 1.2V. The nominal voltages are $V_{RT}=0.00V$ and $V_{RB}=-1.00V$. These voltages may be varied dynamically up to 7MHz. Due to slight variations in the reference current with clock and input signals, R_T and R_B should be low-impedance points. For circuits in which the reference is not varied, a bypass capacitor to ground is recommended. If the reference inputs are varied dynamically as in an Automatic Gain Control (AGC) circuit, a low-impedance reference source is recommended.

Controls

Two function control pins, NMINV and NLINV are provided. These controls are for DC (i.e., steady state) use. They permit the output coding to be either straight binary or offset two's complement, in either true or inverted sense, according to the *Output Coding Table*.

Convert

The TDC1147 uses a CONVert (CONV) input signal to initiate the A/D conversion process. Unlike other flash A/D converters which have a one-clock-cycle pipeline delay between sampling and output data, the TDC1147 requires only a single pulse to perform the entire conversion operation. The analog input is sampled (comparators are latched) within the maximum Sampling Time Offset (t_{STO} , see *Figure 1*). Data from that sample becomes valid after a maximum Output Delay Time (t_O) while data from the previous sample is held at the outputs for a minimum Output Hold Time (t_{HO}). This allows data from the TDC1147 to be acquired by an external register or other circuitry. Note that there are minimum time requirements for the HIGH and LOW portions (t_{PWH} , t_{PWL}) of the CONV waveform and all output timing specifications are measured with respect to the rising edge of CONV.

Analog Input

The TDC1147 uses latching comparators which cause the input impedance to vary slightly with the signal level. For optimal performance, both V_{IN} pins must be used and the source impedance of the driving circuit must be less than 30 Ohms. The input signal will not damage the TDC1147 if it remains within the range of V_{EE} to $+0.5V$. If the input signal is between the V_{RT} and V_{RB} references, the output will be a binary number between 0 and 127 inclusive. A signal outside this range will indicate either full-scale positive or full-scale negative, depending on whether the signal is off-scale in the positive or negative direction.

Functional Description

General Information

The TDC1147 has two functional sections: a comparator array and encoding logic. The comparator array compares the input signal with 127 reference voltages to produce an N-of-127 code (sometimes referred to as a "thermometer" code, as all the comparators referred to voltages more positive than the input signal will be off, and those referred to voltages more negative than the input signal will be on). The encoding logic converts the N-of-127 code into binary or offset two's complement coding, and can invert either output code. This coding function is controlled by DC signals on pins NMINV and NLINV.

Power

The TDC1147 operates from two supply voltages, +5.0V and -5.2V. The return path for I_{CC} (the current drawn from the +5.0V supply) is DGND. The return path for I_{EE} (the current drawn from the -5.2V supply) is AGND. All power and ground pins must be connected.

Reference

The TDC1147 converts analog signals in the range $V_{RB} \leq V_{IN} \leq V_{RT}$ into digital form. V_{RB} (the voltage applied to the pin at the bottom of the reference resistor chain) and V_{RT} (the voltage applied to the pin at the top of the reference resistor chain) should be between +0.1V and -1.1V. V_{RT} should be more positive than V_{RB} within that range. The voltage applied across the reference resistor chain ($V_{RT}-V_{RB}$) must be between 0.8V

Outputs

The outputs of the TDC1147 are TTL compatible, and capable of driving four low-power Schottky TTL (54/74 LS) unit loads. The outputs hold the previous data a minimum time (t_{H0}) after the rising edge of the CONV

signal. New data becomes valid after a maximum time (t_D) after the rising edge of the CONV signal. The use of 2.2 kOhm pull-up resistors is recommended.

Package Interconnections

Signal Type	Signal Name	Function	Value	B7 Package Pins
Power	V _{CC}	Positive Supply Voltage	+5.0V	10, 16
	V _{EE}	Negative Supply Voltage	-5.2V	11, 14
	D _{GND}	Digital Ground	0.0V	4, 21
	A _{GND}	Analog Ground	0.0V	3, 12, 13, 22
Reference	R _T	Reference Resistor (Top)	0.00V	2
	R _B	Reference Resistor (Bottom)	-1.00V	23
Controls	NMINV	Not Most Significant Bit INVert	TTL	5
	NLINV	Not Least Significant Bit INVert	TTL	15
Convert	CONV	Convert	TTL	20
Analog Input	V _{IN}	Analog Signal Input	0V to -1V	1, 24
Outputs	D ₁	MSB Output	TTL	6
	D ₂		TTL	7
	D ₃		TTL	8
	D ₄		TTL	9
	D ₅		TTL	17
	D ₆		TTL	18
	D ₇	LSB Output	TTL	19

TDC1147

Figure 1. Timing Diagram

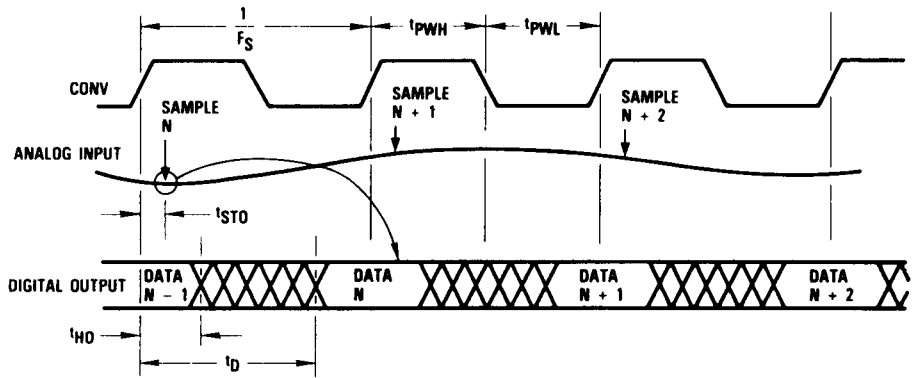


Figure 2. Simplified Analog Input Equivalent Circuit

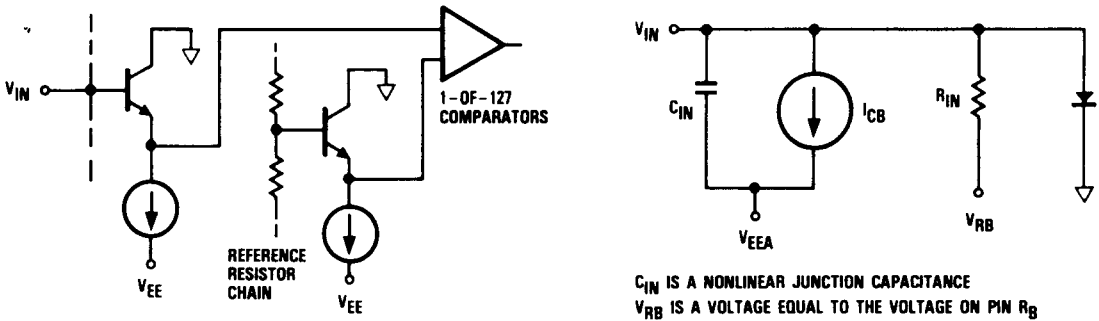


Figure 3. Digital Input Equivalent Circuit

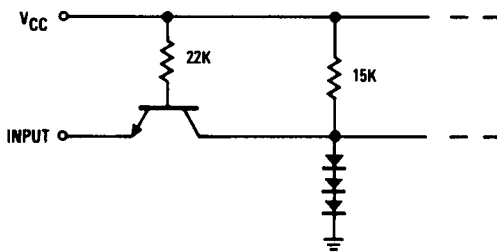
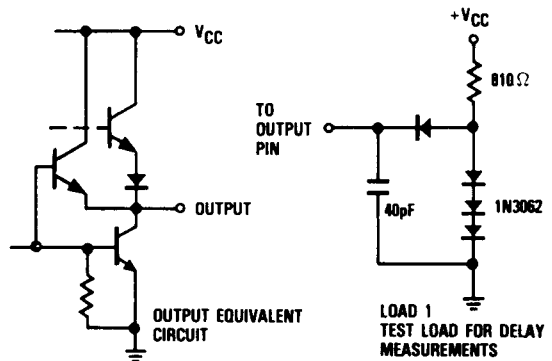


Figure 4. Output Circuits



Absolute maximum ratings (beyond which the device may be damaged)¹

Supply Voltages

V_{CC} (measured to D_{GND})	-0.5 to +7.0V
V_{EE} (measured to A_{GND})	+0.5 to -7.0V
A_{GND} (measured to D_{GND})	-0.5 to +0.5V

Input Voltages

CONV, NMINV, NLINV (measured to D_{GND})	-0.5 to +5.5V
V_{IN} , V_{RT} , V_{RB} (measured to A_{GND})	+0.5 to V_{EE}
V_{RT} (measured to V_{RB})	+2.2 to -2.2V

Output

Applied voltage (measured to D_{GND})	-0.5 to 5.5V ²
Applied current, externally forced	-1.0 to 8.0mA ^{3,4}
Short circuit duration (single output in high state to ground)	1 sec

Temperature

Operating, case	-55 to +125°C
junction	+175°C
Lead, soldering (10 seconds)	+300°C
Storage	-65 to +150°C

Notes:

1. Absolute maximum ratings are limiting values applied individually while all other parameters are within specified operating conditions. Functional operation under any of these conditions is NOT implied.
2. Applied voltage must be current limited to specified range.
3. Forcing voltage must be limited to specified range.
4. Current is specified as positive when flowing into the device.

Operating conditions

Parameter	Test Conditions	Temperature Range						Units
		Standard			Extended			
		Min	Nom	Max	Min	Nom	Max	
V_{CC}	Positive Supply Voltage (measured to D_{GND})	4.75	5.0	5.25	4.5	5.0	5.5	V
V_{EE}	Negative Supply Voltage (measured to A_{GND})	-4.9	-5.2	-5.5	-4.9	-5.2	-5.5	V
V_{AGND}	Analog Ground Voltage (measured to D_{GND})	-0.1	0.0	0.1	-0.1	0.0	0.1	V
t_{PWL}	CONV Pulse Width, LOW	22			22			ns
t_{PWH}	CONV Pulse Width, HIGH	18			18			ns
V_{IL}	Input Voltage, Logic LOW			0.8			0.8	V
V_{IH}	Input Voltage, Logic HIGH	2.0			2.0			V
I_{OL}	Output Current, Logic LOW			4.0			2.0	mA
I_{OH}	Output Current, Logic HIGH			-0.4			-0.4	mA
V_{RT}	Most Positive Reference Input ¹	-0.1	0.0	0.1	-0.1	0.0	0.1	V
V_{RB}	Most Negative Reference Input ¹	-0.9	-1.0	-1.1	-0.9	-1.0	-1.1	V
$V_{RT} - V_{RB}$	Voltage Reference Differential	0.8	1.0	1.2	0.8	1.0	1.2	V
V_{IN}	Input Voltage	V_{RB}		V_{RT}	V_{RB}		V_{RT}	V
T_A	Ambient Temperature, Still Air	0		70				°C
T_C	Case Temperature				-55		125	°C

Note: 1. V_{RT} must be more positive than V_{RB} , and voltage reference differential must be within specified range.

Electrical characteristics within specified operating conditions

Parameter	Test Conditions	Temperature Range				Units
		Standard		Extended		
		Min	Max	Min	Max	
I_{CC} Positive Supply Current	$V_{CC} = \text{Max, static}^1$		25		30	mA
I_{EE} Negative Supply Current	$V_{EE} = \text{Max, static}^1$ $T_A = 0^\circ\text{C to } 70^\circ\text{C}$ $T_A = 70^\circ\text{C}$ $T_C = -55^\circ\text{C to } 125^\circ\text{C}$ $T_C = 125^\circ\text{C}$		-170			mA
			-135			mA
					-220	mA
					-130	mA
I_{REF} Reference Current	$V_{RT}, V_{RB} = \text{Nom}$		35		50	mA
R_{REF} Total Reference Resistance		34		20		Ohms
R_{IN} Input Equivalent Resistance	$V_{RT}, V_{RB} = \text{Nom}, V_{IN} = V_{RB}$	100		40		kOhms
C_{IN} Input Capacitance			60		60	pF
I_{CB} Input Constant Bias Current	$V_{EE} = \text{Max}$		160		300	μA
I_{IL} Input Current, Logic LOW	$V_{CC} = \text{Max}, V_I = 0.5\text{V CONV}$ $NMINV, NINV$		-0.4		-0.6	mA
			-0.6		-0.8	mA
I_{IH} Input Current, Logic HIGH	$V_{CC} = \text{Max}, V_I = 2.4\text{V}$		50		50	μA
I_I Input Current, Max Input Voltage	$V_{CC} = \text{Max}, V_I = 5.5\text{V}$		1.0		1.0	mA
V_{OL} Output Voltage, Logic LOW	$V_{CC} = \text{Min}, I_{OL} = \text{Max}$		0.5		0.5	V
V_{OH} Output Voltage, Logic HIGH	$V_{CC} = \text{Min}, I_{OH} = \text{Max}$	2.4		2.4		V
I_{OS} Short Circuit Output Current	$V_{CC} = \text{Max}, \text{one pin to ground, one second duration.}$		-30		-30	mA
C_I Digital Input Capacitance	$T_A = 25^\circ\text{C}, F = 1\text{MHz}$		15		15	pF

Note:

1 Worst case, all digital inputs and outputs LOW

Switching characteristics within specified operating conditions

Parameter	Test Conditions	Temperature Range				Units
		Standard		Extended		
		Min	Max	Min	Max	
f_S Maximum Conversion Rate	$V_{CC} = \text{Min}, V_{EE} = \text{Min}$	15		15		MSPS
t_{STO} Sampling Time Offset	$V_{CC} = \text{Min}, V_{EE} = \text{Min}$		7		10	ns
t_D Output Delay	$V_{CC} = \text{Min}, V_{EE} = \text{Min}, \text{Load } 1$		60		70	ns
t_{HO} Output Hold Time	$V_{CC} = \text{Max}, V_{EE} = \text{Max}, \text{Load } 1$	15		15		ns

System performance characteristics within specified operating conditions

Parameter	Test Conditions	Temperature Range				Units
		Standard		Extended		
		Min	Max	Min	Max	
E_{LI} Linearity Error, Integral Independent	$V_{RT}, V_{RB} - \text{Nom}$		0.4		0.4	%
E_{LD} Linearity Error, Differential			0.4		0.4	%
CS Code Size	$V_{RT}, V_{RB} - \text{Nom}$	30	170	30	170	% Nominal
V_{QT} Offset Voltage, Top	$V_{IN} - V_{RT}$		+50		+50	mV
V_{QB} Offset Voltage, Bottom	$V_{IN} - V_{RB}$		-30		-30	mV
T_{CO} Temperature Coefficient			± 20		± 20	$\mu V/^\circ C$
BW Bandwidth, Full Power Input		7		7		MHz
t_{TR} Transient Response, Full Scale			10		10	ns
SNR Signal-to-Noise Ratio	7MHz Bandwidth,					
	20MSPS Conversion Rate					
Peak Signal/RMS Noise	1MHz Input	45		46		dB
	7MHz Input	43		44		dB
RMS Signal/RMS Noise	1MHz Input	36		37		dB
	7MHz Input	34		35		dB
E_{AP} Aperture Error			50		50	ps
DP Differential Phase Error ¹	$F_S = 4 \times \text{NTSC}$		1.5		1.5	Degree
DG Differential Gain Error ¹	$F_S = 4 \times \text{NTSC}$		2.5		2.5	%

Note:

1. In excess of quantization.

Output Coding

Range	Binary		Offset Two's Complement	
	True	Inverted	True	Inverted
-1.00V FS	NMINV - 1 NLINV - 1	0 0	0 1	1 0
0.0000V	000000	111111	100000	011111
-0.0078V	000001	111110	100001	011110
•	•	•	•	•
•	•	•	•	•
•	•	•	•	•
-0.4960V	011111	100000	111111	000000
-0.5039V	100000	011111	000000	111111
•	•	•	•	•
•	•	•	•	•
•	•	•	•	•
-0.9921V	111110	000001	011110	100001
-1.0000V	111111	000000	011111	100000

Note:

1. Voltages are code midpoints.

Calibration

To calibrate the TDC1147, adjust V_{RT} and V_{RB} to set the 1st and 127th thresholds to the desired voltages.

Assuming a 0V to $-1V$ input range, continuously strobe the converter with $-0.0039V$ (1/2 LSB from 0V) on the analog input, and adjust V_{RT} for output toggling between codes 00 and 01. Then apply $-0.996V$ (1/2 LSB from $-1V$) and adjust V_{RB} for toggling between codes 126 and 127.

The degree of required adjustment is indicated by the offset voltages, V_{OT} and V_{OB} . Offset voltages are generated by the inherent parasitic resistance between the package pin and the actual resistor chain on the integrated circuit. These parasitic resistors are shown as R_1 and R_2 in the *Functional Block Diagram*. Calibration will cancel all offset voltages, eliminating offset and gain errors.

The above method for calibration requires that both ends of the resistor chain, R_T and R_B , are driven by variable voltage sources. Instead of adjusting V_{RT} , R_T can be connected to analog ground and the 0V end of the range calibrated with an input amplifier offset control. The offset error at the bottom of the resistor chain causes a slight gain error, which can be compensated for by varying the voltage applied to R_B . The bottom

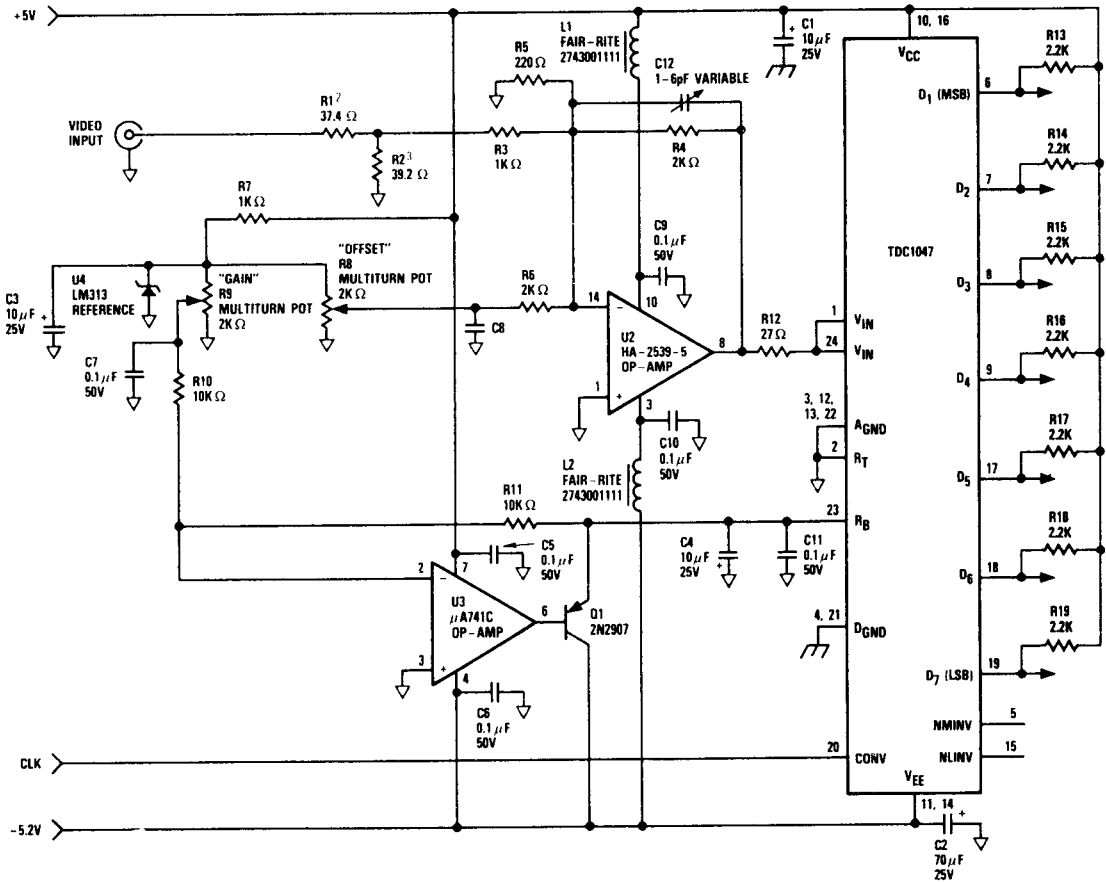
reference is a convenient point for gain adjust that is not in the analog signal path.

Typical Interface Circuit

Figure 5 shows an example of a typical interface circuit for the TDC1147. The analog input amplifier is a bipolar wideband operational amplifier, which is used to directly drive the A/D converter. Bipolar inputs may be accommodated by adjusting the offset control. A zener diode provides a stable reference for both the offset and gain control. The amplifier has a gain of -1 providing the recommended $1V_p$ input for the A/D converter. Proper decoupling is recommended for all supplies, although the degree of decoupling shown may not be needed. A variable capacitor permits either step response or frequency response optimization. This may be replaced with a fixed capacitor, whose value depends upon the circuit board layout and desired optimization.

The bottom reference voltage, V_{RB} , is supplied by an inverting amplifier, followed with a PNP transistor. The transistor provides a low-impedance source and is necessary to sink the current flowing through the reference resistor chain. The bottom reference voltage can be adjusted to cancel the gain error introduced by the offset voltage, V_{OB} , as discussed in the *Calibration* section.

Figure 5. Typical Interface Circuit



Notes:

1. Unless otherwise specified, all resistors are 1/4W, 2%.

$$2. R1 = Z_{IN} \left(\frac{1000 R2}{1000 + R2} \right)$$

$$3. R2 = \frac{1}{\left(\frac{ZV_{Range}}{V_{REF} Z_{IN}} \right) - 0.001}$$

TDC1147

Ordering Information

Product Number	Temperature Range	Screening	Package	Package Marking
TDC1147B7C	STD - $T_A = 0^\circ\text{C}$ to 70°C	Commercial	24 Pin CERDIP	1147B7C
TDC1147B7V	EXT - $T_C = -55^\circ\text{C}$ to 125°C	MIL-STD-883	24 Pin CERDIP	1147B7V